

# Circuit phase-locks function generators over 360°

by Lawrence W. Shacklette  
Seton Hall University, South Orange, N. J.

This circuit locks a low-cost function generator without a voltage-controlled-oscillator (VCO) input to a second inexpensive generator having such a phase-reference feature. A J-K flip-flop, two one-shots, two comparators, and an operational amplifier are linked in a feedback arrangement that includes the programmable generator as the VCO in a phase-locked loop (PLL). The resultant circuit provides a selectable phase shift between generator outputs over the range of 0° to 360°.

In operation, the output of the low-cost generator,  $v_1$ , passes through a high-pass filter to a zero-crossing detector that employs a 311 comparator ( $M_1$ ), as shown. A dual monostable multivibrator and J-K flip-flop follow. Although these two devices can be eliminated, they enable the phase-locked loop to be operated at the center of its locking range for any phase shift. This arrangement ensures that two often desired phase angles, 0° and 180°, fall within the capture range of the PLL, so that if locking is lost, it will be automatically regained.

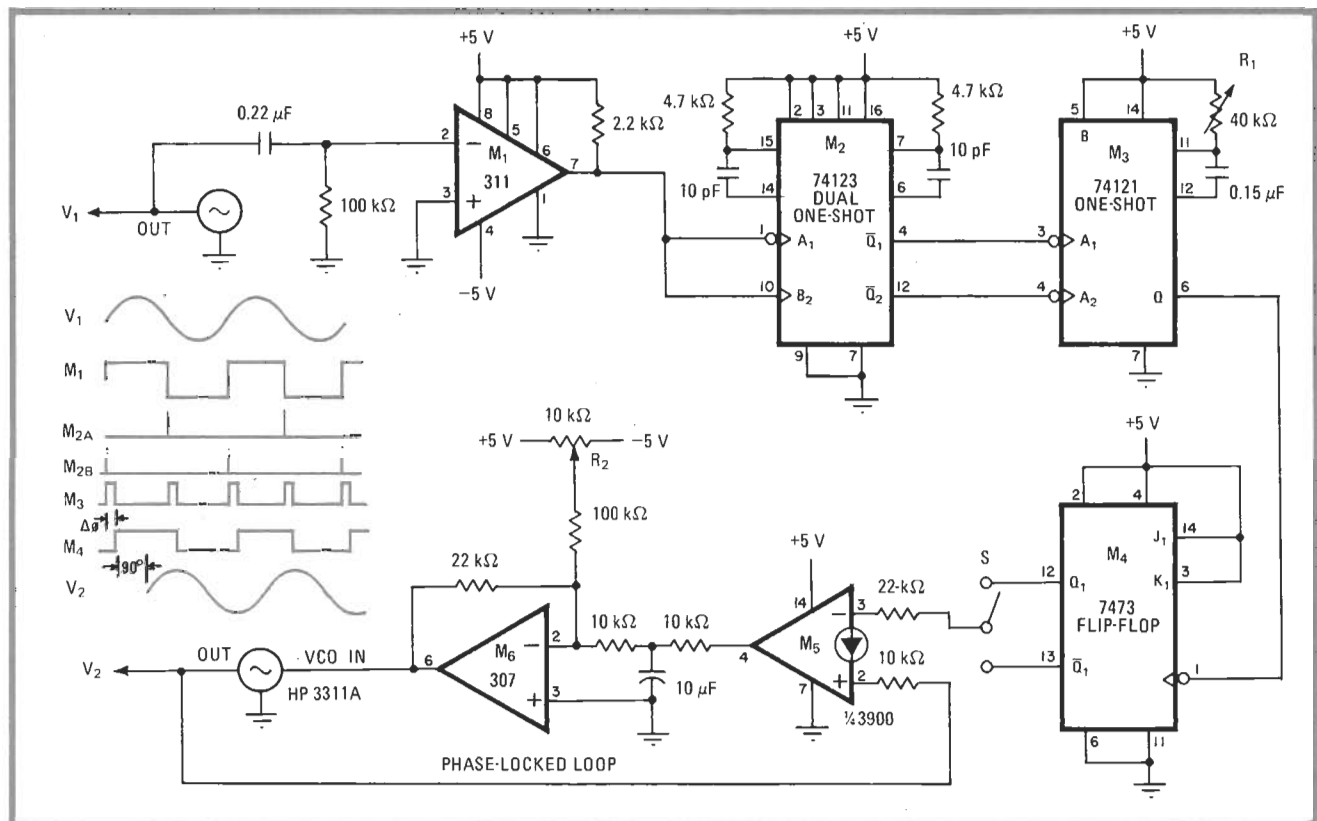
The outputs of the dual one-shot,  $M_2$ , wired so that each fires on opposite edges of the signal applied to its

inputs (see timing diagram), are fed to the OR input of  $M_3$ , whose on time is selected by potentiometer  $R_1$ . Thus  $R_1$  controls the amount of phase shift.

$M_3$  produces two pulses for each cycle of  $v_1$  and triggers the J-K flip-flop,  $M_4$ , on each negative edge. The flip-flop thus produces a square wave with a frequency equal to  $v_1$ , but shifted in phase by up to 180°. An additional shift of 180° can be obtained by using switch  $S$  to connect the Q output of  $M_4$  to the inverting input of the 3900 Norton amplifier.

The adjustable-phase square wave serves as a reference signal for the phase-locked loop, which is composed of the 3900, a low-pass filter, a buffer (307), and a VCO (the second function generator, a Hewlett-Packard 3311A). The input signal to the VCO is a negative dc voltage that is the inverted sum of the filtered output of the 3900 and the voltage selected by the offset control,  $R_2$ . By turning the generator's front-panel control or  $R_2$ , the free-running frequency of the loop can be adjusted.

Because the reference signal is a square wave, the PLL will lock onto either the fundamental of  $v_1$  or its odd harmonics. Selection of a particular harmonic is made by adjusting the free-running frequency to the approximate value of the harmonic desired. Half-multiple harmonics ( $\frac{1}{2}f_1$ ,  $\frac{3}{2}f_1$ ,  $\frac{5}{2}f_1$ , etc.) can be produced at  $v_2$  by breaking the  $\bar{Q}_1$ - $A_1$  connection between  $M_2$  and  $M_3$  and tying  $A_1$  to +5 volts. Even harmonics can be obtained by using the remaining flip-flop in the 7473 as a divide-by-2 counter, and placing it between the output of the VCO and the 3900's noninverting input. □



**Phase-locked.** Comparators, one-shots, and flip-flop combine to provide stable locking of generators without phase-reference feature to those having a VCO input.  $R_1$  and  $S$  are used to select phase of  $v_2$  with respect to  $v_1$ ; phase can be adjusted from 0° to 360°.  $M_2$  and  $M_4$  ensure that locking is regained if it is lost, and  $R_2$  controls lock frequency, which may be set to integer or half-integer harmonics of  $v_1$ .