## Use a switching-regulator controller to generate fast pulses

Mitchell Lee, Linear Technology Corp



and fall times into a  $50\Omega$  load.

A source of pulses with fast-rising edges that approximate the step function can help you perform many useful laboratory measurements, including characterization of coaxial cables' rise times and location of cable faults using time-domainreflectometry methods. For example, evaluating the rise time of a 10- to 20ft-long RG-58/U cable requires edgetransition times of 1 to 2 nsec. Agilent's (www.agilent.com) HP8012B, a workhorse pulse generator that finds use in many electronics labs, can deliver pulses with rise times of 5 nsec that are adequate for many applications but not for cable characterization.

As an alternative, switching-regulator-controller ICs can deliver gatedrive pulses with rise and fall times of less than 2 nsec, making them ideal candidates for laboratory pulse-generation service. A simple implementation uses Linear Technology's (www. linear.com) LTC3803 constant-frequency flyback controller,  $IC_1$  (**Figure** 1). The controller self-clocks at 200 kHz, and applying a sample of its output to its Sense pin causes the controller to operate at its minimum duty cycle and produce a 300-nsec-wide output pulse.

The LTC3803's output can deliver more than 180 mA into a 50 $\Omega$  load. so use a low-series-inductance bypass capacitor that connects as directly as possible between IC<sub>1</sub>'s power and ground (pins 5 and 2). The decoupling components,  $C_1$ , a 10- $\mu$ F ceramic capacitor, and  $R_1$ , a 200 $\Omega$  resistor, minimize pulse-top aberrations without introducing amplitude droop. The circuit's output directly drives a 50 $\Omega$  termination at amplitudes as high as 9V. For applications that require maximum pulse fidelity, use a backtermination resistor,  $R_{BACKTERM}$ , to suppress triple-transit echos and absorb reflections from the cable and any mismatch in the cable's far-end termination impedance. Back-termination also helps when driving passive filters, which expect to see a specific generator impedance. The LTC-3803's output impedance is approximately  $1.5\Omega$ , which affects the value of the back-termination resistor. The back-termination technique (continued on pg 106)

## designideas

works well with load impedances of at least 2 k $\Omega$ . At impedances higher than that value, parasitic impedances associated with the terminating resistor and IC<sub>1</sub> degrade bandwidth and pulse fidelity.

In a back-terminated,  $50\Omega$  system,

the circuit delivers a 4.5V output pulse with symmetric rise and fall times of 1.5 nsec, pulse-top-amplitude aberrations of less than 10%, and amplitude droop of less than 5%. Directly driving a  $50\Omega$  load doesn't degrade the output's rise and fall times. For best

pulse fidelity, use stripline techniques to route  $IC_1$ 's output directly to the termination resistor and output connector  $J_1$ . Using a 100-mil-wide trace on a  $^1/_{16}$ -in., double-sided, glass-epoxy pc board approximates a 50 $\Omega$  surge impedance.EDN