

Switching regulator efficiently controls white-LED current

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A few years ago, manufacturers specified their white, but dim, LEDs for a maximum forward-current rating of 20 mA. Today's white LEDs deliver more light and thus must operate at ever-higher bias currents. Maintaining control of an LED's bias point while operating at high current near its maximum rating requires a new approach.

The simplest and most common method of biasing an LED involves connecting a resistor in series with the LED to limit the LED's maximum current, but this method directly impacts power efficiency, which you define as the ratio of power to the LED to the total input power. For a white LED operating at 350 mA, the corresponding forward-voltage drop across the diode is approximately 3.2V. A series resistor and LED connected to a 5V power source operates at 64% efficiency—that is, 3.2V for a 5V source. The power dissipates as heat, causing an average power loss in the series resistor of 36 mW at a forward current of 20 mA, which is acceptable, but this figure balloons to 630 mW at a forward current of 350 mA.

In addition, using a series resistor allows the diode's bias point and thus its brightness to fluctuate as the

power-supply voltage and the ambient temperature vary. Based on National Semiconductor's (www.natsemi.com) LM2852 switched-mode bucking regulator, which features internal compensation and synchronous-MOSFET switches that can drive loads as large as 2A, the circuit efficiently provides constant-current drive to a high-current LED and minimizes the effects of supply-voltage and temperature variations on the LED's brightness (**Figure 1**).

In this circuit, the LM2852 operates at efficiency of approximately 93% and directly controls a step-down-regulator topology that maintains a constant current flow through LED₁, which

potentiometer R₁ adjusts. Current-to-voltage conversion taking place within the circuit's control loop effectively regulates the circuit's output current. In operation, the LM2852 compares its internal reference voltage with the voltage from the divider formed by D₁, R₁, and R₂ and drives the control loop to maintain a constant 1.2V at its voltage-sense pin. Current through the voltage divider is proportional to the current through LED₁, and the ratio of the currents tracks over the circuit's operating-temperature range because D₁ and LED₁ exhibit approximately the same forward-voltage temperature coefficient of $-2 \text{ mV}/^\circ\text{C}$. Mounting D₁ and LED₁ next to each other on the pc board provides sufficiently close thermal coupling for temperature compensation.

With R₁'s wiper fully clockwise, the

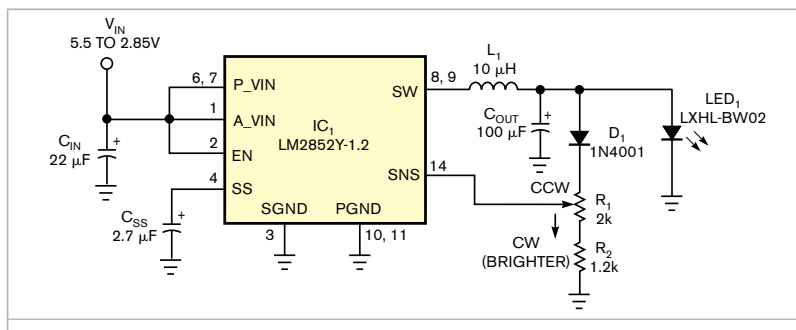


Figure 1 This circuit drives a high-current, white LED at 93% efficiency over input voltage and temperature. Potentiometer R₁ controls current through LED₁ and allows brightness adjustment. Diode D₁ provides temperature compensation for LED₁'s forward-voltage drop.

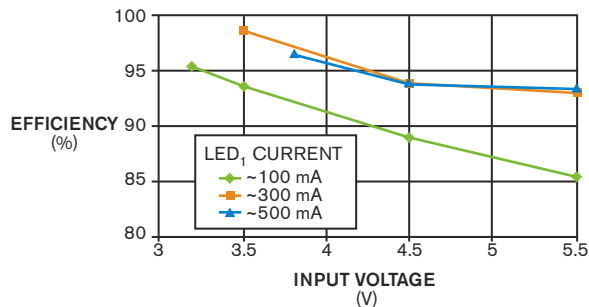


Figure 2 Circuit efficiency versus input voltage shows an increase in efficiency for increasing LED current and decreasing input voltage.

current through D_1 approaches 1 mA, and the current through LED₁ averages approximately 500 mA. Adjusting R_1 counterclockwise reduces LED₁'s forward current from 500 mA to 0A.

When scaling the values of R_1 and R_2 for a different current-loop gain, decreasing the gain impacts the circuit's conversion efficiency, and increasing the gain makes the loop more sensitive to component tolerances. To provide a remote brightness control, you can replace mechanical potentiometer R_1 with a digitally programmed potentiometer. Luxeon (www.luxeon.com), the manufacturer of LED₁, an LXHL-

BW02, specifies limits of 350-mA continuous current and 500-mA peak-pulsed current. **Figure 2** shows the circuit's efficiency versus variations in input voltage. Note that the circuit's efficiency increases as input voltage decreases, which helps extend operating time in battery-powered-system applications.

As temperature fluctuates, the current through LED₁ varies less than 3% over the temperature range, a factor-of-three improvement over a series-resistor current-limiting circuit (**Figure 3**). Although more complex than a single resistor, the circuit in **Figure 1** requires

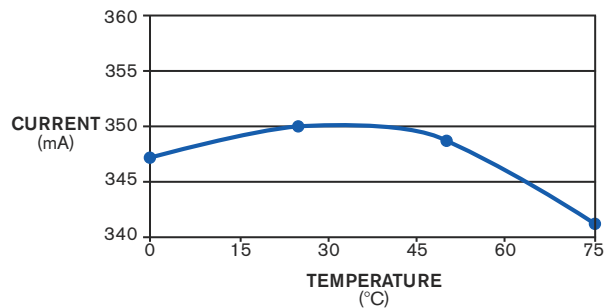


Figure 3 Current through the LED varies less than 3% over an operating-temperature range of 0 to 75°C.

only a few components. For L_1 , this prototype uses Coilcraft's (www.coilcraft.com) MSS5131-103 surface-mount inductor rated for 10 μ H.

National Semiconductor's data sheet for the LM2852 outlines criteria for selecting capacitors C_{IN} , C_{SS} , and C_{OUT} . For efficient heat removal, the circuit's pc board should include generous copper-mounting pads and traces for IC₁ and LED₁. At a forward current of 350 mA, LED₁ dissipates 1.1W, so consult the manufacturer's data sheet to review its thermal-design recommendations. **EDN**