

# Inexpensive logic controls stepper motor

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A number of sophisticated ICs for stepper-motor control are now available. However, the advanced features of these chips—self-clocking, high-current drive, and full-step, half-step, and direction control—are often unnecessary or remain unused. For a design that needs to control only the number of steps, drive speed, and direction, you can make a very simple and inexpensive driver using two low-level logic chips (**Figure 1**). The cost of this controller is less than \$1; the cost of dedicated motor-control ICs starts at around \$5. The drawback is a slight increase in board space.

Going back to the basics, you can control a standard stepper-motor drive, whether bipolar or unipolar, using a four-step sequence (**Table 1a**). By replacing the on and off states with ones and zeros, respectively (**Table 1b**), Column B becomes the logical inverse of Column A, and Column D becomes the logical inverse of Column C. Thus, the corresponding state diagram (**Figure 2**) comprises just 2 bits. Clockwise rotation results from using a logical one to move sequentially from state one to state four and back to state one. Likewise, counterclockwise rotation results from using a logical zero to move through the states in the reverse order.

TABLE 1—STEPPER-MOTOR-DRIVE SEQUENCE

State	A	B	C	D	State	A	B	C	D
1	ON	OFF	ON	OFF	1	1	0	1	0
2	ON	OFF	OFF	ON	2	1	0	0	1
3	OFF	ON	OFF	ON	3	0	1	0	1
4	OFF	ON	ON	OFF	4	0	1	1	0

(a) CCW (upward arrow) (b) CW (downward arrow)

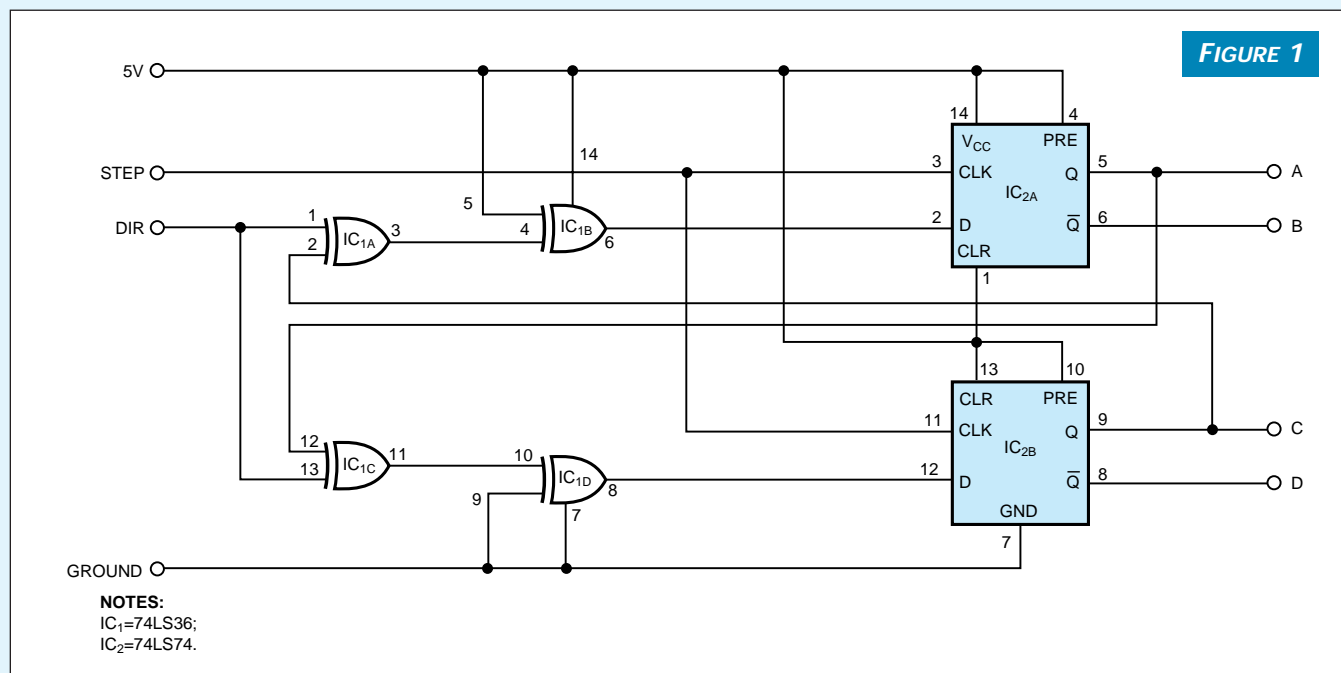
You can then produce the present-state/next-state assignment (**Table 2**) and the next-state maps (**Figure 3**). Then, by inspection, the logical choice is to loop the state maps out for D flip-flops, which produces the following two logic equations:

$$DA = ((DIR)(C)) + ((DIR)(\bar{C})); \quad (1)$$

$$DB = ((DIR)(A)) + ((DIR)(\bar{A})). \quad (2)$$

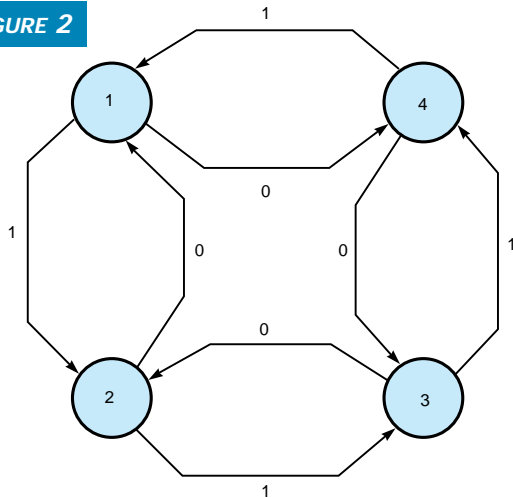
**Equation 1** is an exclusive NOR, and **Equation 2** is an exclusive OR. To save space, you can use a single quad XOR chip to implement both equations. A dual D flip-flop completes the logic driver, as **Figure 1** shows. Using rising-edge-triggered D flip-flops helps keep the design simple while eliminating mode-change faults.

The circuit derives the four outputs from the Q and  $\bar{Q}$



Two logic-level ICs can implement simple and inexpensive control of a stepper-motor driver.

FIGURE 2

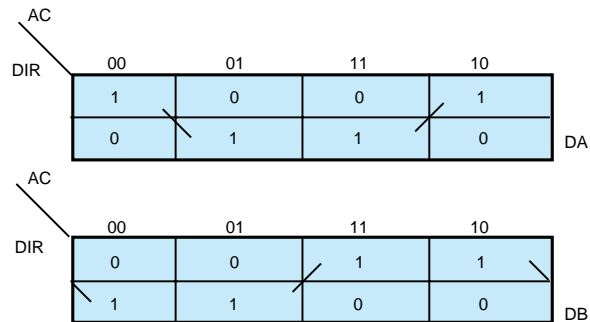


The state diagram for the stepper-motor controller comprises just 2 bits.

outputs of D flip-flops  $IC_{2A}$  and  $IC_{2B}$  in **Figure 1**.  $IC_{1A}$  XORs the Q output of flip-flop  $IC_{2B}$  with the DIR input, and the circuit transforms the output into an XNOR by using  $IC_{1B}$  as a controlled inverter.  $IC_{1B}$  then drives the D input of flip-flop  $IC_{2A}$ . Similarly,  $IC_{1C}$  XORs the Q output of  $IC_{2A}$  with the DIR input. The output of  $IC_{1C}$  drives XOR  $IC_{1D}$ , which acts as a noninverting buffer. The output of  $IC_{1D}$  drives the D input of  $IC_{2B}$ . Using XOR gate  $IC_{1D}$  as a buffer keeps the propagation delays to the D inputs of the flip-flops equal, which helps the circuit avoid any race conditions. The STEP signal is the step-rate input, which drives the clock inputs of both flip-flops.

The last design task is to add the appropriate-sized transistors to drive the stepper motor. In the case of the unipolar motor, output signals A, B, C, and D can directly drive the transistors. To drive a bipolar motor, you can use the A and C outputs to drive one-half of two H-bridges and the B and D outputs to drive the other corresponding half of the

FIGURE 3



The next-state maps correspond to two simple logic equations.

**TABLE 2—PRESENT-STATE/NEXT-STATE ASSIGNMENT TABLE**

Present state			Next state	
A	C	DIR	A	C
0	0	0	1	0
0	0	1	0	1
0	1	0	0	0
0	1	1	1	1
1	0	0	1	1
1	0	1	0	0
1	1	0	0	1
1	1	1	1	0

H-bridges. This design is possible because the B output is the inverse of A, and D is the inverse of C. (DI #2176)

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