

# Letters to the Editor

---



---

## THE LANGUAGE OF HI-FI

---



---

Your balanced and sensible leader in the August issue came as balm to my inflamed spleen after also reading in one of your considerably less distinguished contemporaries that a highly respected preamplifier "sounded boring" and "made the music sound as if played by amateurs". Surely the nadir of lunacy in the use of subjective language! One gets the impression that these terminological outrages are being perpetrated on gullible readers by a new breed of journalistic *wunderkind*, who would probably be hard pressed to define a decibel. The reasons for this development are beyond me — probably it is either an effort to conceal technical incompetence or because it makes saleable copy; or a mixture of both.

Of course, I am not against the use of subjective language. What I am against is the increasing tendency to use language of imprecise meaning. To misquote Gertrude Stein "a volt is a volt" and I hope no one is going to question that or challenge that a volt measured in hi-fi equipment is any different from any other. But when someone says *vis-à-vis* the performance that the "information retrieval efficiency was low" (yes, really — I didn't make it up) then like the late and quite unlamented Hermann Goering, I reach for my axe. If I as an experienced professional engineer cannot understand it, then heaven help the poor layman.

We commentators in engineering journalism have a heavy responsibility and should never resort to language that is capable of alternative interpretation or is open to doubt; and if there is a slight doubt, then it should be clearly defined or explained. At the risk of being accused of pedantry, I will go further and say that every observed phenomenon in reproduced sound is measurable and may be expressed in quantitative terms. Some subtle effects perhaps may be harder to measure than others; but I am with Galileo and Lord Kelvin. Inventing new words is not the way out.

May I finish with another observation, and a warning against another tendency not confined to the popular hi-fi press? This is the lack of a sense of proportion and a failure to appreciate the realities of the technical side of audio. I have just been reading with interest an article in a well-known technical publication. The writer discusses with great insight, the technical desiderata for a pickup input stage; then spoils it all by proudly declaiming in the final paragraphs that the

improvements result in a reduction of the t.h.d. to 0.0004%. Marvellous. Then if someone is able to make a gramophone record and cartridge capable of the same order of inherent  $Dt$  we might just be able to notice the difference.

Reg Williamson  
Norwich

---



---

## AURAL SENSITIVITY TO PHASE

---



---

I fear that Mr Moir (Letters, July 1977 issue) has misunderstood the point which I was trying to make in my letter on the audibility of polarity reversals (Letters, May 1977). Far from the distortion of one stage in the amplifier chain being cancelled by a complementary distortion in a subsequent stage, as suggested by Mr Moir as an explanation for the effects I discussed, I was at pains in my letter to make clear that this was *not the case*. All subsequent stages in the chain, including the transducer, were shown not to be responsible for the effect in question. (In the case of the loudspeaker, this was done by listening from both front and back of the dipolar electrostatic panels, thus introducing a polarity reversal in the acoustic waveform, which was found to reverse the effect.) The change in quality of the signal was due entirely to its own asymmetry, not to subsequent distortion. This confirms the earlier work cited in my letter.

An even more vivid demonstration of this effect can be obtained by linearly combining two sinusoidal oscillator signals, one a "fundamental" frequency of around 400Hz and the other an adjustable-level "second harmonic" of around 800Hz. If the second harmonic is allowed to drift slowly in phase relative to the fundamental a very pronounced cyclic change in the sound quality of the signal will be heard, and it is instructive to listen to it while observing the asymmetric waveform on an oscilloscope. No such effect appears to occur if the 800Hz signal is shifted to the third harmonic, i.e. 1200Hz; the waveform is now always symmetric with respect to polarity reversals. With a fourth harmonic, however, the effect is again subtly audible if the level is suitably chosen.

Towards the end of his letter, Mr Moir in fact seems to support my argument, by agreeing that *on good signals* a polarity reversal is indeed subtly audible. This strikes me as being an important conclusion! Even more than just standardizing the absolute polarity of the whole audio chain, as I suggested, it would seem that the non-linear-phase errors inherent in the use of pressure and/or velocity microphones in recordings, which are reproduced indiscriminately via either pressure or velocity transducers, also requires serious investigation.

Stanley P. Lipshitz,  
University of Waterloo,  
Ontario, Canada.

Mr Driscoll, responding in the July issue to my letter of last February, asserts of himself "My grasp of basic principles is not so uncertain that I could believe Coleman's claim that "tone bursts which differ in the framing of phase" (I wrote "OR 'phase'") of the sine wave with respect to the burst envelope have spectra of different shapes." My claim can easily be checked, and is

correct. Where does that leave his "grasp of basic principles"?

If the members of a regular sequence of tone bursts are well separated, so that they are heard as separate bursts, it is enough to calculate the Fourier transform or spectrum of any one of them. If a particular burst consists of the sinusoid  $\sin(2\pi f_0 t + \epsilon)$  gated on for  $2n$  periods centred about the time  $t=0$  then its transform is

$$K \sqrt{(f-f_0)^{-2} + (f+f_0)^{-2} + (f^2 - f_0^2) \cos 2\epsilon} \sin(2\pi n f / f_0) e^{j\phi(f)}$$

where  $\phi(f) = \epsilon - \tan^{-1} \left( \frac{f-f_0}{f+f_0} \cos 2\epsilon \right) + \pi/2$  and  $K$  is independent of both  $f$  and  $\epsilon$ . If the burst is not a whole number of periods long the expression becomes more complicated.

This spectrum peaks at  $f=f_0$  and the width of the peak, taken between neighbouring zeros, is  $f_0/n$ , inversely proportional to the burst length, and compatible with the requirements of the acoustic uncertainty relationship. Its shape, i.e. the variation of its modulus with  $f$ , clearly does change when the value of  $\epsilon$  changes, and in addition the reference phase  $\phi(f)$  of the component of frequency  $f$  depends in a non-linear fashion on both  $f$  and  $\epsilon$ . If the centre of the burst occurs, not at time  $t=0$ , but at  $t=T$ , then  $\phi(f)$  contains a further additive term  $-2\pi f T$ . If  $\epsilon = \pi/2$  the spectrum of the burst decays at frequencies far from  $f_0$  as  $f^{-1}$ , whereas if  $\epsilon = 0$  it decays as  $f^{-2}$ . This is understandable since in the latter case the burst has discontinuities of slope at its ends, but in the former has amplitude discontinuities, which will splash the spectrum out much further, a point about which I warned Mr Driscoll in my February letter. He doesn't have to take my word for these statements — presumably one of his brighter students could check the calculations, or he could ask one of the enterprising loudspeaker manufacturers who have set themselves up with minicomputers, f.f.t. programmes, and graphics terminals to let him see for himself what a sine wave toneburst spectrum really looks like, in phase as well as in amplitude.

It is all too easy for those acquainted in principle with Fourier transforms to mention the use of transfer functions and Fourier transforms for calculating network responses to signals of finite duration, leaving the impression that this is essentially a trivial extension of normal a.c. calculations. It is not, and exposure to the specific Fourier transforms of a few simple signals, such as tone bursts, can go a long way towards driving the point home.

C. F. Coleman,  
Wantage,  
Oxon.

---



---

## CONFUSION ABOUT DISTORTION?

---



---

In a letter in your August issue Mr Greenbank quotes an earlier correspondent who states: "... 'loss of information' occurs during amplifier 'latch-up' — when, as we all know, 100% intermodulation distortion occurs." This statement is symptomatic of a general confusion which has resulted from harmonic distortion, intermodulation distortion, "latch-up", "clipping", "slew-rate limiting", and transient intermodulation distortion all being regarded as "distortion".

The use of distortion as a generic term is probably responsible for it being generally unnoticed that the above list may be the

results produced by two fundamentally differing mechanisms.

Consider the case of an amplifier which, though it has a non-linear transfer function, has no clipping point or slew-rate limit. Such an amplifier may be modelled by a "one-to-one" mapping function, and because of this an inverse mapping function may be discovered which precisely restores any mapped set of points back to their initial positions. With any distortion which may be described this way, therefore, we always (in principle, at least) perform another process which gives us the information in its "undistorted" form.

Such is not the case with "latch-up", "clipping" and "slew-rate limiting". Each of these may not be regarded as a "one-to-one" mapping — rather, they are characterised by a "many-to-one" mapping function. In these cases no inverse mapping function exists which may be employed to restore any arbitrary initial point to its original position. We have created a singularity, and a set of points are "doomed to fall down it".

For this reason it will unfortunately tend to cloud the issue to regard "many-to-one" imperfections in a transfer function as "distortion". Hence it is misleading to regard clipping or latch-up as "100% intermodulation distortion". Similarly, it is unhelpful to call the effects of slew-rate limiting "transient intermodulation distortion."

I would not wish to argue that "many-to-one" imperfections are not "distortion" as the word is currently defined — only that we are here clouding the problem by our choice of terms.

As for the "loss of information" concept which prompts Mr Greenbank's letter, all I can do is point out that this may be defined in terms of "many-to-one" rather than "one-to-one" functions. It remains to be seen, however, if either form of imperfection proves inherently "audibly more objectionable".

J. C. G. Lesurf,  
Armstrong Audio Ltd,  
London N7.

## THE E.M. EQUATIONS — ALTERNATIVE REPRESENTATION

Maxwell's equations relating the electromagnetic field to charge and current densities are usually presented in vector form:

$$\nabla \cdot D = \rho \tag{i}$$

$$\nabla \times H = J + \frac{\partial D}{\partial t} \tag{ii}$$

$$\nabla \cdot B = 0 \tag{iii}$$

$$\nabla \times E = -\frac{\partial B}{\partial t} \tag{iv}$$

Tensor formulation of Maxwell's equations is even more concise and expresses better the interdependence of electric and magnetic fields.<sup>1,2,3</sup>

An alternative method of representing equations (i) and (ii) is shown in Fig. 1. Starting from six components of  $D$  and  $H$  (circled symbols) we operate on them as indicated by the direction of arrows. We then obtain three components of the current

Fig. 1

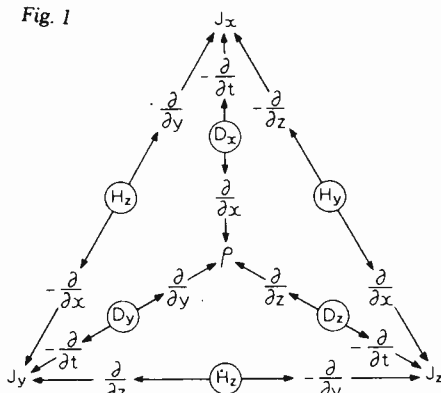
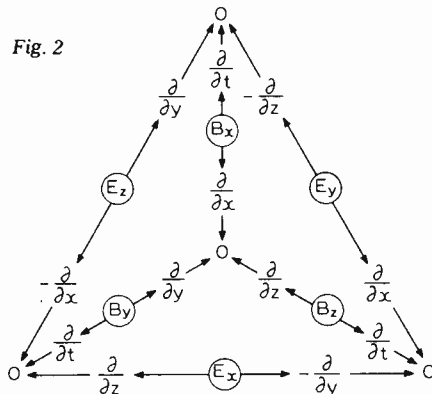


Fig. 2



density in the apices of the triangle and charge density at the middle.

In the same way we can express the remaining two Maxwell's equations (iii) and (iv), as shown in Fig. 2.

Using the same method we can represent the relations between electric and magnetic fields and the four-vector potential ( $V, A_x, A_y, A_z$ ) viz. the equations:

$$B = \nabla \times A \tag{v}$$

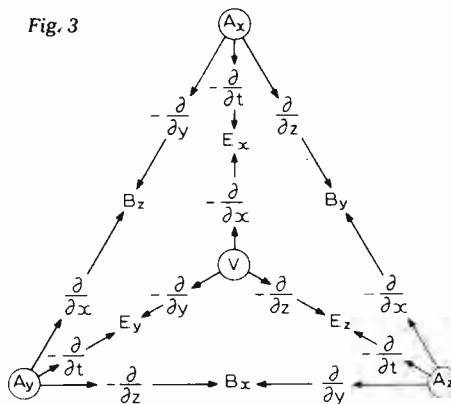
$$E = -\nabla V - \frac{\partial A}{\partial t} \tag{vi}$$

The alternative representation of the above equations is shown in Fig. 3.

Here we start from the components of the four-vector potential (circled symbols in the middle and at the apices of the triangle), and operate on them as indicated. We obtain the six components of  $E$  and  $B$ .

The advantage of the above representation of four differential equations, relating six vector components of the electromagnetic field and four components of the four-vector current or four-vector potential, is mainly

Fig. 3



mnemonic, but it also helps to grasp the essential unity of electric and magnetic fields.

T. A. Kasinski,  
Kingston-upon-Thames,  
Surrey

### References

1. A. Lichnerowicz, "Elements of Tensor Calculus" pp. 143-150, Methuen & Co. Ltd., London, 1962
2. A. N. Matveyev, "Principles of Electrodynamics" pp. 329-337, Reinhold Publishing Corporation, New York, 1966.
3. G. Taylor, "Special Relativity" pp. 65-71, Clarendon Press, Oxford, 1975

## COMPUTING FOR LOCAL COMMUNITIES

We want to discover if it's truly possible to introduce computing into one of London's most derelict areas as a community resource.

We are the Vauxhall Media Project and our primary aim is to initiate a meeting place and facilities for film, video, photography, printing and computing. These are facilities which groups and individuals from the local community can use to fulfil their own projects and the needs of their community by pooling talent and resources.

The story so far is that a group of people working in the computer industry and in the local community have been meeting regularly to analyse the type of system needed.

John Pemberton of London University has been very helpful with advice and there is a probability that computer time may be found there to run a graphic computer terminal five nights a week.

We want to see if computing can be brought into the community as a tool and as a medium for creative entertainment, and we would welcome information, advice, help, participation, equipment — anything.

Peter Fotheringham,  
Vauxhall Media Project  
132 South Lambeth Road,  
London SW8.

## INCONSIDERATE TV CAMERA OPERATION

One in seven persons in the United Kingdom suffers from migraine. In addition, a considerable number are subject to epilepsy. All such people in their capacity as television viewers are badly affected by flashing lights, rapidly rotating symbols and most kinds of unsteady image thrown at them from the television screen.

As these facts are well known, why do the BBC and the IBA continue to allow their producers and cameramen to indulge in flashing light techniques, to "hosepipe" their lenses, and — worst of all — to pump zoom lenses back and forth?

Significantly, these are the first pitfalls that a beginner in movie photography is taught to avoid. This in itself should be enough to justify their discontinuance. But their dire effect on very many viewers of the small