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Other Parts Discussed in Post: [DAC8830](#), [OPA340](#), [OPA388](#)

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In applications such as position sensors, data-acquisition systems and resistance temperature detectors (RTDs), it is important to design with high precision in mind. In many cases, designing with precision integrated circuits (ICs) reduces signal-chain complexity, lowers the external component count, and minimizes board space and bill of materials (BOM) costs. The inaccuracies of one device may propagate through with the inaccuracies of another device, resulting in undesirable and unpredictable errors. In the case of a buffer-configured [operational amplifier](#) (op amp) at the output of a [digital-to-analog converter](#) (DAC), it's crucial that your DAC and your op amp are precision devices for an accurate output.

A traditional rail-to-rail complementary metal-oxide semiconductor (CMOS) amplifier architecture includes two differential pairs, PMOS (blue) and NMOS (red), shown in Figure 1. Together, these two transistor pairs span the entire input common-mode voltage range. When one transistor pair takes over from the other, however, a unique and nonlinear phenomenon known as “input crossover distortion” occurs due to the intrinsic input offset voltage of each of the two input differential pairs, shown in Figure 2.

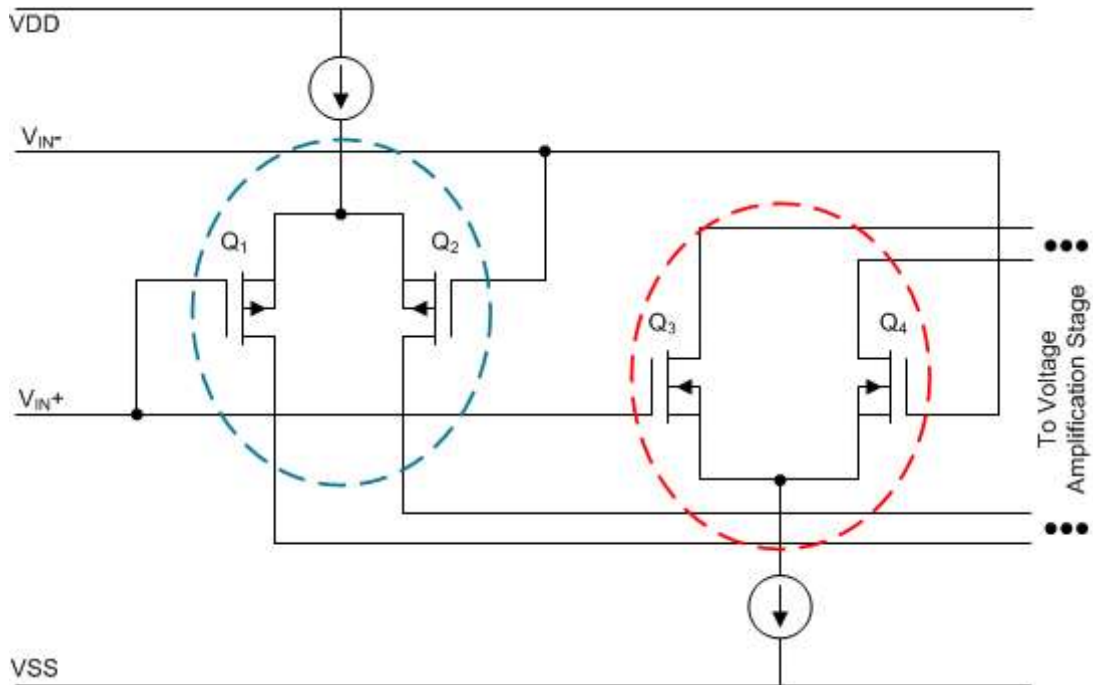


Figure 1: Traditional rail-to-rail CMOS amplifier architecture

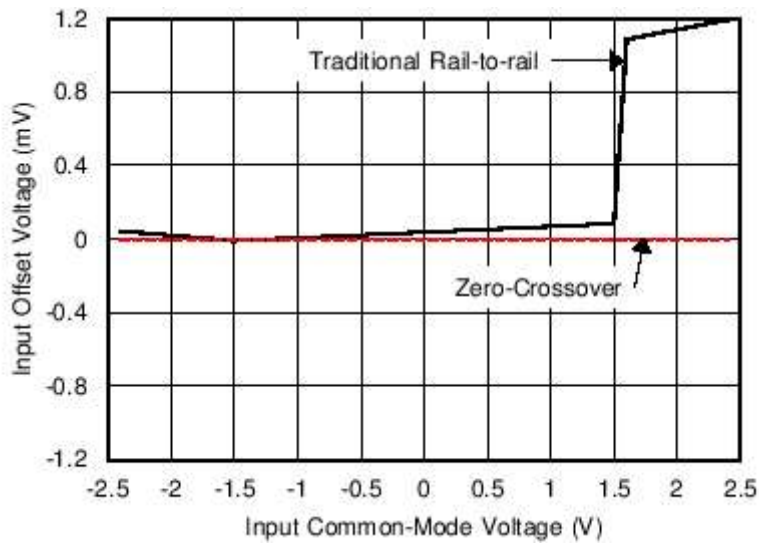


Figure 2: Input offset voltage vs. common-mode voltage

When you connect a traditional rail-to-rail CMOS op amp at the output of a high-precision DAC, the crossover distortion will introduce an error and result in a drastic increase in integral nonlinearity (INL). This may cause the signal to deviate several least significant bits (LSBs) from its ideal value.

Now, what does 1LSB mean? Equation 1 is a simple equation to calculate LSB:

$$1LSB = \frac{V_{ref}}{2^N} \quad (1)$$

- where N is the DAC's number of bits.

The DAC8830 is a 16-bit DAC. If the voltage reference is $V_{ref} = 5V$, then:

$$1LSB = \frac{5V}{2^{16}} = 76.3\mu V \quad (2)$$

So to deviate more than 1LSB means that you can have more than 76.3 μV of error at your output. This can be detrimental to many precision applications, like critical systems where failure has the potential to negatively impact customers' end products.

So how do you fix this? Enter zero crossover!

You can span the entire input common-mode voltage range by using a zero-crossover op amp such as the [OPA388](#). The zero-crossover topology uses an internal regulated voltage charge pump to increase the positive supply voltage and thus achieve linear operation with input common-mode voltages all the way to its rails with a single input transistor pair, shown in Figure 3. This results in true rail-to-rail input operation without a crossover region, and thus no crossover distortion. If you were to connect this kind of op amp at the output of a DAC, the op amp does not introduce an error within the common-mode region (1V to 2V below the positive rail) like a traditional rail-to-rail CMOS device.

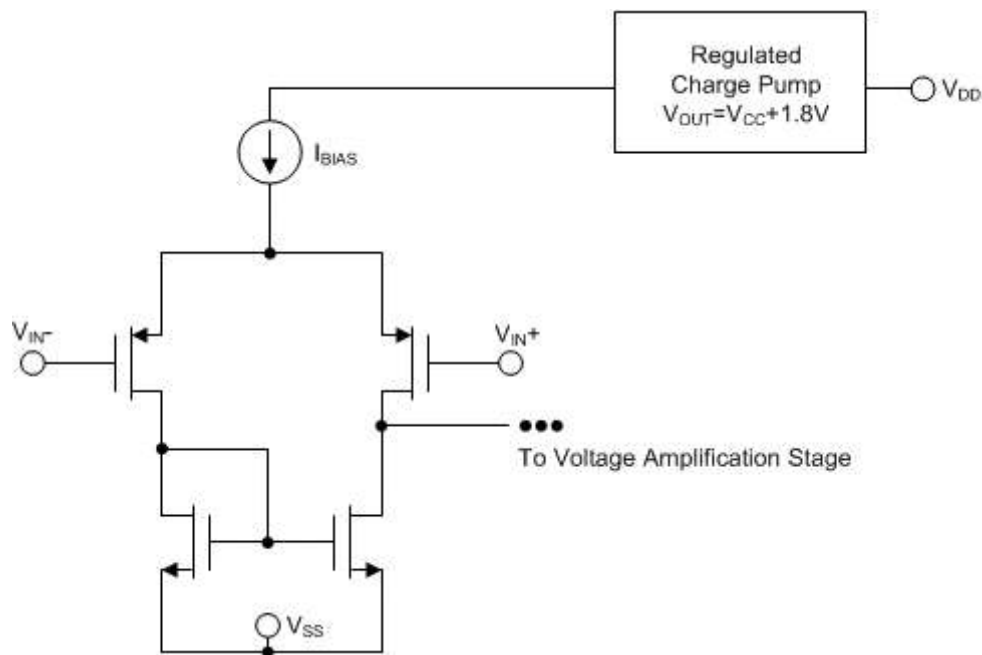


Figure 3: Zero-crossover amplifier architecture

In Figure 4, the black curve describes the output of a traditional rail-to-rail CMOS op amp (OPA340) at the output of a DAC ([DAC8830](#)), while the red curve describes the output of a [zero-crossover op amp](#)

- (OPA388) with the same DAC8830. As you can see, the output of the DAC8830 + OPA388 does not suffer from the distortion that is easily visible in the DAC8830 + OPA340 output curve. The [High-Precision Reference Design for Buffering a DAC Signal](#) describes this output in greater detail.

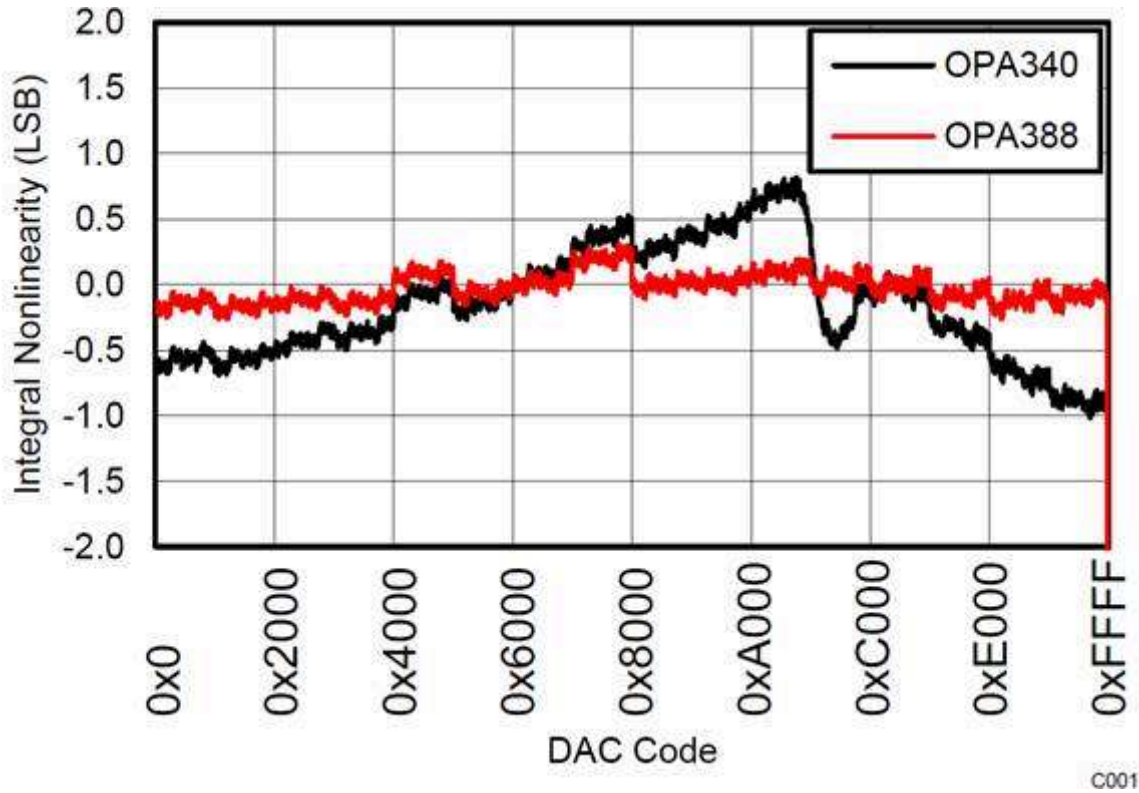


Figure 4: INL comparison of rail-to-rail CMOS OPA340 and zero-crossover OPA388

Let's put this reference design into perspective and use it in an application such as an MRI machine. An MRI uses a powerful magnetic field to produce detailed 2-D and 3-D pictures of the human body to diagnose and/or monitor several health conditions. Unacceptably distorted signals that exceed the error budget in any way can potentially impair the quality of the images.

The OPA388 is the industry's first op amp to employ zero-crossover and zero-drift technology. Zero-drift op amps have an internal self-correcting circuit that produces ultra-low input offset voltage (V_{OS}) and near-zero input offset voltage drift over time and temperature (dV_{OS}/dT). The technology also delivers other advantages, including no $1/f$ noise (flicker noise), low broadband noise (white noise) and low output distortion, which can help increase system reliability in harsh environments. Take a swimming pool for instance – pH pool testers and monitoring systems must withstand changes in the environment's temperature to correctly determine the deficit or excess of chlorine. Since most pools are placed outside, the environment's temperature can vary many degrees between a cold winter's day and a hot summer's day. Offset voltage will change with temperature deviations, introducing error, so it is crucial to select an op amp with low offset voltage drift to support system reliability through these changes.

- To assure high performance, high precision and high accuracy, carefully select parts for your design. Make sure that you understand your system and what you can afford in terms of error, and only then sift through [TI's diverse portfolio](#) for your ultimate solution.

Additional resources

- Download the [zero-drift](#) and [zero-crossover](#) Tech Notes.
- Visit the [OPA388](#) and [DAC8830](#) datasheets to learn more.

 2 comments  0 members are here



[Jia ChenQiang](#) *over 2 years ago*

Hi,Where to download the zero-drift technical manual in Additional resources?



[Kuba Sunderland-Ober](#) *over 1 year ago*

Input crossover distortion is only a problem when the common mode is not held constant. Best performance from any op-amp, be it brand new design or decades old, usually requires the common mode to be constant. So, these op-amps solve a problem that doesn't come up if the application design was done right to start with. Alas, sometimes non-inverting stages seem to be the "simplest" solution. And then the complexity shifts onto the IC op-amp designer :) If you don't want to marry the design to this unique part, and don't have a spare charge pump laying around, it's best to make the whole thing insensitive to both CMRR and crossover distortion in one fell swoop.