

# Lowpass filter has improved step response

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A COMMON PROBLEM that arises when you design lowpass filters for signal conditioning is the filters' effect on the system's time-domain response. Because pushing the cutoff frequency lower slows the step response, the system may fail to recognize significant changes within a reasonable amount of time. The circuit in **Figure 1** accommodates lower cutoff frequencies without sacrificing the step-response time. A window comparator monitors the delta (difference) between the filter's input and output. When the delta exceeds  $\pm 50$  mV, the filter increases its slew rate by increasing the cutoff frequency by an order of magnitude. The switched-capacitor filter, IC<sub>1</sub>, normally operates as a self-clocked device. Capacitors C<sub>1</sub> and C<sub>2</sub> set the cutoff frequency at 0.1 Hz, and other circuitry forms a dynamic window comparator. Transistor pairs Q<sub>1</sub>-Q<sub>2</sub> and Q<sub>3</sub>-Q<sub>4</sub> form a complementary current mirror whose output flows through R<sub>2</sub> and R<sub>3</sub>, creating a delta of  $\pm 50$  mV. Connecting the output voltage to the center tap of the two resistors centers the delta on the output voltage. You therefore set the window comparator's

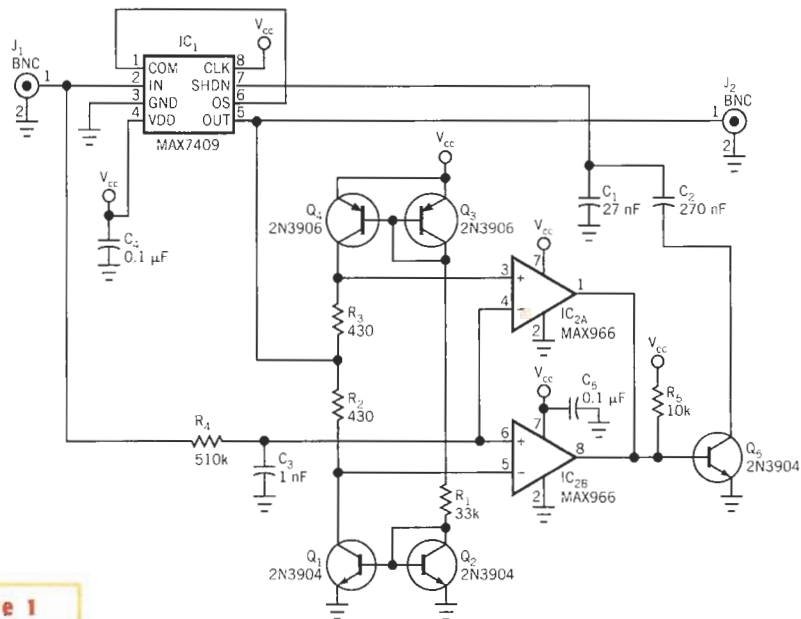


Figure 1

This lowpass filter maintains a fast step response by dynamically adjusting its cutoff frequency.

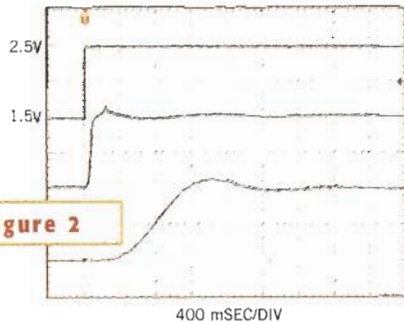
upper threshold at  $V_{OUT} + 50$  mV and the lower threshold at  $V_{OUT} - 50$  mV.

R<sub>4</sub> and C<sub>3</sub> provide lowpass-filtering to the original input signal, producing a

312-Hz cutoff frequency that reduces sensitivity to momentary glitches. The filtered input drives the window comparator's input. If that input is outside the

$\pm 50$ -mV window, comparator IC<sub>2A</sub> or IC<sub>2B</sub> asserts its output low. The low output drives Q<sub>5</sub> into cutoff, causing its collector to assume a high impedance. Because Q<sub>5</sub>'s collector no longer grounds capacitor C<sub>2</sub>, the filter's cutoff frequency increases by a factor of 10. When the system's output changes to within 50 mV of the input, the cutoff frequency throttles back to its quiescent state. **Figure 2**'s oscilloscope photo shows the effect. The top trace is a step from 1.5 to 2.5V, the middle trace is the output with optimization circuitry enabled, and the bot-

tom trace shows the filter's unmodified response. The optimized response includes a slight perturbation during the cutoff-frequency transition, but is five times faster than that of the unmodified circuit. The circuit in **Figure 1** is configured for low cutoff frequencies, but you can rescale it for higher frequencies by changing C<sub>1</sub> and C<sub>2</sub>. You can also modify R<sub>2</sub> and R<sub>3</sub> for different window values, for which the delta equals the resistance multiplied by 115  $\mu$ A. The comparator must be an open-drain type. □



**Figure 2**

These traces show the time-domain response for the circuit in **Figure 1** with optimization circuitry (middle trace) and without it (bottom trace).