

# Analog-input circuit serves any microcontroller

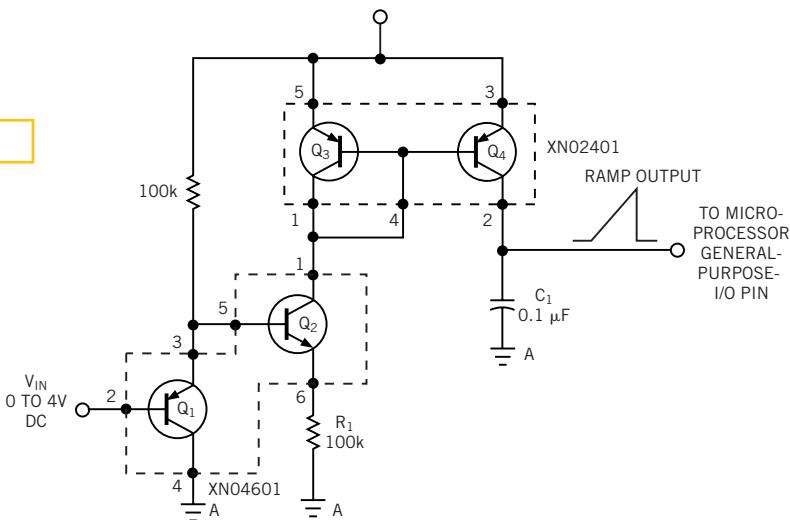
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**T**HE SIMPLE ADC in **Figure 1** is perfect for getting analog signals into a purely digital microcontroller. Using just five surface-mount parts, you can assemble it for less than 50 cents (1000), which is ap-

**Figure 1**

proximately half the cost of a single-chip-ADC approach in the same volume. Moreover, this design takes only one pin from the microcontroller to operate. Although you can purchase many microcontrollers with built-in ADCs, in some circumstances, this solution is impractical. For example, you might have an all-digital microcontroller already designed in. In this design, a USB-compatible, digital-only microcontroller needed analog input at low cost for a consumer application. The basic analog portion of the circuit in **Figure 1** uses clever transistor arrays from Panasonic (www.panasonic.com).  $Q_1/Q_2$  and  $Q_3/Q_4$  are single-package, multiple-transistor arrays. The  $Q_1/Q_2$  array forms a voltage-to-current converter. The voltage on  $Q_1$ 's emitter is a diode drop higher than the voltage on  $Q_1$ 's base. The  $V_{BE}$  drop in  $Q_2$  returns the original input voltage to the top of  $R_1$ ;  $R_1$  then converts that voltage to a current.

The  $Q_3/Q_4$  array forms a standard current-mirror circuit. The current flowing in  $Q_3$ 's collector matches the current forced in  $Q_4$ 's collector.  $Q_4$ 's collector has high impedance, so  $Q_4$  provides a suitable current source. The current from  $Q_4$  charges  $C_1$  at a rate that is proportional to the input voltage. The values in **Figure 1** allow for a range of conversion times of 3 msec for an input of 4V to 56 msec for an input of 0.1V. The design exploits the fact that most general-purpose microcontrollers have a bidirectional I/O-port structure. That is, you can program a port pin as either an input or an output. When you set a pin as an input, it has very high input impedance, so it can follow the ramp as  $C_1$  charges up. When you program a pin as an output, you can set it low, and it discharges  $C_1$  for the next conversion cycle. This action gives you the basic operation of a single-slope analog-to-digital-conversion cycle.



**With two transistor arrays and three discrete components, you can configure an analog front end for a microcontroller.**

The basic operations are as follows:

1. Set the ADC pin as a low output to discharge  $C_1$ .
2. Reset a suitable timer-counter in the microcontroller.
3. Set the ADC pin as an input.
4. Allow the timer to count until it reads as logic 1 in the microcontroller, or let the timer count to some suitably long value, which suggests that the input is essentially zero.
5. Stop the timer counter.
6. Convert to the timer count by some suitable scaling factor to an ADC reading.
7. Start over for the next conversion.

The conversion from the ramp time to a logic 1 on the microcontroller pin depends on the following factors:

- the logic-1 switching level of your microcontroller;
- the input voltage and, hence, the ramp rate of  $C_1$ ;
- the value of  $C_1$ , which sets the ramp rate;
- the value of  $R_1$ , which sets the ramp rate; and
- the microcontroller's timer resolution.

You can boil down these variables to the following equation:

$$\frac{C_1 V_L}{dT} K = V_{IN},$$

where  $V_L$  is the voltage level of the microcontroller's zero-to-one conversion,  $K$  is the scaling factor that relates to the voltage-to-current conversion of the input stage and timer resolution, and  $dT$  is the time count of the conversion cycle. Because  $C_1 V_L$  is also a constant for a given circuit, you can combine it with  $K$  to form a single conversion constant of  $K_1$ . Hence, you can reduce the equation to  $K_1/dT = V_{IN}$ .

In this case, the test code was written for Microchip Technology's (www.microchip.com) PIC16F84 microcontroller. This device has a measured  $V_L$  of 1.28V; the counter has a resolution of 1  $\mu$ sec. It's probably best to empirically determine the factor  $K_1$ . Set up the counter resolution as desired, allow the microcontroller to make and display that conversion time or send it through a debugger, and, given that you have an exact  $V_{IN}$ ,  $K_1$  is then easy to determine. In this case,  $K_1$  turned out to be  $2V \times 5700 \mu\text{sec} = 11,400$ .

The constant  $K_1$  serves to convert the raw timer count to a voltage. To obtain high resolution, you normally use floating-point math. If you need to display the value, floating-point math might be ap-



## design ideas

appropriate, but most applications entail reading a potentiometer or some other system level. In such applications, the output is a bar-chart display or some control value. Thus, you waste microcontroller resources by using floating-point math throughout the conversion process. With careful selection of circuit components, fixed-point math can usually provide, for example, an 8-bit repre-

sentation (0 to 255) for an input range of 0 to 4V. If you scale the timer/counter by 64, instead of a count of 5700  $\mu$ sec for an input of 2V, you obtain 89. Then, if you want this 89 to correspond to a half-scale value of 128, the value of  $K_1$  becomes 11,392. A 16-bit unsigned word easily accommodates this value, and you need no floating-point math in the conversion. The accuracy of this ADC is ap-

proximately 5% with no adjustments. The resolution is a function of the timer resolution and how tight the code makes the conversion loop. The resolution can be many times the absolute accuracy. Moreover, the converter is monotonic.

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